

# Quantifying Timing Skew In Differential Signaling using Practical Fiber Weave Model

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# Outline

### Practical Fiber Weave Effect Model for Skew and Jitter

- Introduction
- Modeling fiber-weave effect with non-uniform transmission line segments
- Printed circuit board test vehicle
- Model identification with loosely coupled traces
- Model identification and measurement validation with tightly coupled traces
- Conclusion

### Isola Low-Skew Product Solutions

- Tachyon<sup>TM</sup>100G
- GigaSync<sup>TM</sup>
- Chronon<sup>TM</sup>



# **Skew Modeling Methods**

Many Skew Modeling Methods are Available with Variable Levels of Complexity and Accuracy Brute Force Fiber Weave Model



mproved Accuracy

Unit Cell Cascaded in System Tool

Model Complexity/Solve Time

# Introduction

- Communication data links on PCBs are running at bitrates of 10-30 Gbps and beyond
  - Design of interconnects for such links is a challenging problem that requires electromagnetic analysis with causal material models from DC to 20-50 GHz
- Woven fabric composites are typically used as insulators to manufacture PCBs
- Both fabric fiber and resin are composite materials have different dielectric constant (DK) and loss tangent (LT) properties:

Typical Dielectric Material Property	DK	DF	Differential Glass Weave V1 V1 V2 V2
Glass Weave	4.4 - 6.1	0.002 - 0.007	
Resin	3.2	0.003 - 0.027	Impregnated Resi

- Dielectric inhomogeneity in transmission line cross-section causes mode conversion or skew
- Inhomogeneity along the line causes resonances in insertion and reflection losses
- Both effects may contribute to deterministic jitter and have to be modelled and mitigated if necessary
- A practical fiber-weave effect model is presented



### **Model for Non-uniform Dielectric Across Traces**

We use the **Imbalance Factor** to characterize dielectric properties variation (specified with Imbalance as shown on the right);

**Unit Imbalance Factor** corresponds to volume average resin percentage defined for the given PCB material globally;

Variation upwards corresponds to higher volumetric content of glass (higher dielectric constant and smaller polarization losses);

Variation downward corresponds to higher volumetric content of the resin (smaller dielectric constant and larger polarization losses);

### Quasi-static field solver is used to build such model





### **Model for Non-uniform Dielectric Along Traces**

**Modulation Factor is used** to characterize dielectric property variation (specified either with step values as shown on the right or with periodic functions of length);

**Unit Modulation Factor** corresponds to volume average resin percentage defined for the given PCB material globally;

Variation upwards corresponds to higher volumetric content of glass (higher dielectric constant and smaller polarization losses);

Variation downward corresponds to higher volumetric content of the resin (smaller dielectric constant and larger polarization losses);



#### Concatenation of t-line segments with adjusted dielectric properties is used to model this effect



### Causal Model for Variable Dielectric Properties Option 1

Apply product of Imbalance and Modulation Factors to dielectric constant at infinity (causal adjustment):

Multi-pole Debye model:  $\varepsilon(f) = \phi \cdot \varepsilon(\infty) + \sum_{n=1}^{N} \frac{\Delta \varepsilon_n}{1 + i \frac{f}{fr}}$ 

Wideband Debye model (aka Djordjevic-Sarkar ):

$$\varepsilon_{wd}(f) = \phi \cdot \varepsilon(\infty) + \varepsilon_{rd} \cdot F_d(f)$$

$$F_d(f) = \frac{1}{(m_2 - m_1) \cdot \ln(10)} \cdot \ln\left[\frac{10^{m_2} + if}{10^{m_1} + if}\right]$$

Other causal models can be adjusted similarly

- $\phi$  = ImbalanceFactor · ModulationFactor
- $\phi = 1$  corresponds to the original "homogenized" model;
- $\phi > 1$  increases the dielectric constant at infinity and automatically decreases the loss tangent;
- $\phi < 1$  decreases the dielectric constant at infinity and automatically increases the loss tangent;



### Causal Model for Variable Dielectric Properties Option 2

Apply product of Imbalance and Modulation Factors to volume fraction in mixing formulas (also causal):

Wiener upper boundary model (layered dielectric):

 $\varepsilon_{eff,\max} = \phi \cdot f \cdot \varepsilon_2 + (1 - \phi \cdot f) \cdot \varepsilon_1$ 

Wiener lower boundary model (comb-like dielectric):

$$\varepsilon_{eff,\min} = \frac{\varepsilon_1 \cdot \varepsilon_2}{\phi \cdot f \cdot \varepsilon_1 + (1 - \phi \cdot f) \cdot \varepsilon_2}$$

Hashin-Shtrikman and Maxwell-Garnett models can be adjusted similarly

- $\phi$  = ImbalanceFactor · ModulationFactor
- $\phi = 1$  corresponds to the original "homogenized" model;
- $\phi > 1$  increases the dielectric constant and automatically decreases the loss tangent;
- $\phi < 1$  decreases the dielectric constant and automatically increases the loss tangent;

Assuming dielectric 2 is glass with higher DK and lower LT, dielectric 1 is resin with lower DK and higher LT and both simulated with causal models



### Test Board for Numerical Experiments and Experimental Validation

#### Test Board Stackup to investigate 2 materials from Isola

				Material : GigaSync/I-SPEED			
	Turpe	Structure (Stack up)	Cu weight	Construction	Thickness after	DK/DF	
LTR	туре		(oz)		iam (mii)		
	ImAg Finish						
1	TOP		0.5 + plating		2.1		
	prepreg			Gigasync 2116 - RC 60%	5.0	4.13/.0067	
2	GND		0.5		0.6		
	core			Gigasync 2116	4.5	4.13/.0066	
3	S3		0.5		0.6		
	prepreg			Gigasync 2116 - RC 60%	4.4	4.13/.0067	
4	GND		0.5		0.6		
	core			I-SPEED 3X1652	19.0	3.72/.007	
5	GND		0.5		0.6		
	prepreg			I-SPEED 3313 - RC 61.5%	4.4	3.50/.007	
6	S6		0.5		0.6		
_	core			I-SPEED 3313	4.0	3.65/.007	
7	GND		0.5		0.6		
	prepreg			I-SPEED 3313 - RC 61 5%	4.8	3.50/.007	
8	BOT		0.5 + plating		21	0.001.001	
	ImAg Finish		pitting				
				Pressed thickness	53.9		

Gigasync: Wideband Debye model because of glass and resin have close DK

I-SPEED: Wiener average mixture of S-glass with Dk=5 and LT=0.001 and 61.5% resin with Dk=2.8 and LT=0.011 @ 1 GHz (produces Dk=3.5, LT=0.007 as in specifications)



6-inch microstrip differential links with probe launches on top (GigaSync 2116) and bottom (I-SPEED 3313) of the board;



# **Two Cases Considered**

- Loosely coupled pairs: trace width 9 mil, separation 39.5 mil (Kv=0.012, center to center 9.7+2\*19.4 mil)
- Tightly coupled pairs: trace width 4.9 mil, separation
  4.8 (Kv=0.21, center to center 9.7 mil)





### **De-compositional Model of Test Structure**



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# Model identification for worst case skew (numerical example)

From: L. Ritchey, J. Zsio, R. Pangier, G. Partida, "High speed signal path losses as related to PCB laminate type and copper roughness", DesignCon 2013.

TEST PCB SKEW D								
			VERTICAL 9"				14"	
MATERIAL	WEAVE	MINIMUM	MAXIMUM	AVERAGE		MINIMUM	MAXIMUM	AVERAGE
IS415	3313	0	8	5		30	123	88
FR408HR	3313	1	8	5		3	43	20
FR408HRIS	8313	0	7	4.6		6	20	11.8
I-SPEED	3313	3	10	4.5		1	59	18
I-SPEED LOW DK	8313	1	4	2.3		5	12	7.5
I-TERA	3313	1	12	6		1	13	9.5
I-TERA LOW DK	8313	1	4	2.5		4	59	24.6

Worst case observed on I-SPEED with 3313 glass style in un-coupled traces is 59 ps or 4.2 ps/inch

1. Use 5 ps/inch as the maximal possible skew due to FWE and adjust the **Imbalance Factor** for loosely coupled line to observe the same skew;

2. Estimate jitter due to skew in loosely coupled lines;

3. Define **Modulation Factor** along the line with the same amplitude as the imbalance and see effect on jitter;



### **Identification of Imbalance for Worst Case Skew**

Imbalance = 0.2 (Imbalance Factor 0.9/1.1 or resin content +/-10%) produces skew 5 ps/inch in loosely coupled differential pair



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# Impact of the worst case imbalance on insertion loss and mode transformation (loosely coupled traces)

Differential to common mode transformation is zero if no imbalance; Very large far end mode transformation observed with Imbalance 0.2 (+- 10% of resin); Mode transformation also degrades differential insertion loss (IL);



Mode transformation is shown to be good way to quantify imbalance effect on PCB composite dielectric



# Impact of worst case skew on jitter (loosely coupled traces)

#### 25 Gbps PRBS 7 signal, 10 ps rise and fall time



Substantial reduction of eye width (timing jitter) and eye height is expected For case of 10% imbalance – worse case for I-Speed with 3313 glass weave



# Impact of +-10% resin content variation along the line (loosely coupled traces)

#### Strips are running at 7 degree to horizontal fiber – no imbalance, maximal modulation period 164 mil, amplitude 0.2 (+/-10% variation of the resin content) 25 Gbps, PRBS 7, 10 ps rise and fall – no



No substantial effect on jitter expected (due to narrow band of the resonance)



## **Test Board for Tightly Coupled Traces**

#### This board was manufactured, simulated and

investigated experimentally

5 microstrip structures with offset for I-SPEED/3313 on the bottom side;

5 microstrip structures with offset for Gigasync/2116 on the top side;

TDR measurements are done by Brian Butler from Introbotix;

S-parameter measurements are done by Reydezel Torres Torres from INAOEP;

Analysis with Simbeor software;





# Direct TDR measurements for tightly coupled traces

 Worst case for MS1 - about 6 ps (1 ps per inch) produces Imbalance = 0.05 (Imbalance Factor 0.975/1.025)



TDR measurements and simulation are done with all ports open Tight coupling case is much less sensitive to material variations



## S-parameters, Tightly Coupled Traces

### 6-in links on I-SPEED/3313:



Mode transformation is smaller than expected from the TDR measurements – the imbalance is closer to 0.025 (+,- 1.25% resin variation) versus assumption of 0.05

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# Imbalance Impact on Eye Diagram

#### 6-in links on I-SPEED/3313; Signal: 25 Gbps, PRBS 7, 10 ps rise time;



Analysis with Balanced strips produce practically the same eye (no visible difference); Why simulated and measured eyes are slightly different? – see next slide...



# Why eyes are slightly different?

6-in links on I-SPEED/3313; single-ended TDR computed from measured S-parameters with Gaussian step 16 ps rise time



These effects are more considerable than investigated imbalance of 0.05?



## S-parameters, tightly coupled traces

### 6-in links on GigaSync/2116:



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### **Conclusion: Fiber-Weave Effect (FWE) modelling**

- New causal non-uniform imbalanced transmission line model for prediction of FWE on signal propagation in PCB interconnects has been introduced
  - Imbalance Factor is used for inhomogeneity across traces
  - Modulation Factor is used for inhomogeneity along traces
  - Both factors are applied either to DK at infinity for simple dispersive models or to volume fraction in two dielectric mixture formulas
- Model parameters can be identified with either worst case skew or worst case far end mode transformation (diff. to common)
- Usability of the models are illustrated with examples of practical investigation of corner cases for I-SPEED and GigaSync dielectrics (<u>www.isola-group.com/products</u>)
- Proposed models are implemented in Simbeor software (<u>www.simberian.com</u>)



### Conclusion: Fiber-Weave Effect (FWE) and Jitter

- FWE impact on jitter and eye height for a 25 Gbps signal were evaluated:
  - Numerical experiments conducted for loosely coupled pairs
  - Numerical and experimental investigations for tightly coupled pairs
- Significant effect of imbalance on jitter for loosely coupled microstrip pairs has been observed
- Little effect of periodicity observed on jitter for loosely coupled pairs is observed
- No significant effect of imbalance on tightly coupled microstrip traces
- Consistent tightly-coupled traces are not easily achieved in practice, so design considerations revert to loosely coupled case





# Isola Low-Skew Product Solutions





# Tachyon<sup>®</sup> 100G

Ultra Low Loss, Leadfree Laminates and Prepregs for HSD Applications

### **Tachyon® 100G Value Proposition**

- Engineered To Improve Insertion Loss on the Most Demanding High Speed Digital Designs
- Tachyon-100G is recommended For 40+ Gb/s Backplane and Line Cards
- Constructions have been Optimized to Improve CAF and Lead-free Assembly Performance
- Complete Line of Laminates And Prepregs With Spread Glass Weaves To Minimize Micro-Dk Effects Of Glass Fabrics And to Mitigate Skew
- HDI Design Friendly
- Can be Used in Hybrid Builds as Prepregs and Laminates Because of the Low Cure Lamination Cycle







# Chronon®

### Next Generation Low Loss, Low Skew & Leadfree Laminates and Prepregs for HSD Applications

# **Chronon Value Proposition**

- Engineered To Eliminate Skew Issues In Differential Pairs On High Data Rate Designs
- Targeted For 40+ Gb/s Designs (Backplanes And Line Cards) That Require More Bandwidth
- Optimized Constructions To Improve Lead-free Assembly Performance
- Offer Laminates And Prepregs With Engineered Glass Weaves To Minimize Micro-Dk Effects Of Glass Fabrics And to Mitigate Skew
- Eliminates the Need To Rotate Circuitry on The Laminates
- UL Approved In Same Family As I-Tera®MT and Chronon® Simplifying UL Recognition Process PCB Fabricators





# High-Speed Digital Products Lee Ritchey SI TV

### Speeding Edge Signal Integrity Test Vehicles



#### **Courtesy of Speeding Edge**



### **16 Layer SI TV Stackup**

						1.100	SS / 1		
			Copper	Material	Material	Material			
			Type S =	Pressed Er	Unpressed	Pressed		Copper	Copper
Laver	Material		RTF, $X =$	(at ~2	Thickness	Thickness		Thickness	Thickness
#	Type	Material Construction	HVLP	GHz)	(mils)	(mils)	Picture	(mils)	(oz)
				- /	( - /	< - /			
						0.7	Solder Mask		
1						011		22	1.5
	Prepred	1 x 3313 RC = 57%		3.68	4.3	4 1	Proprog		
2	1.001.09		x	0.00				0.6	0.5
-	Core	1 x 3313 RC - 55%	core	3 72		А		0.0	0.0
3	0010		X	0.72				0.6	0.5
0	Prepred	2 x 3313 PC - 57%	X	3.68	8.6	83	Broprog	0.0	0.0
1	Fiepleg	2 x 3313 KC = 37 /8	v	3.00	0.0	0.5	Prepreg	0.6	0.5
4	Coro	1 x 2212 PC - 55%	<b>A</b>	2 7 2		1		0.0	0.5
Б	Core	1 X 3313 RC = 35%		3.72		4	Core	0.6	0.5
Э	Dreverser	2 × 2212 DC 57%	^	2.00	0.0	0.0	5	0.6	0.5
0	Prepreg	$2 \times 3313 \text{ RC} = 57\%$	v	3.68	8.6	8.3	Prepred	0.0	0.5
6			X	0.70			6	0.6	0.5
	Core	$1 \times 3313 \text{ RC} = 55\%$	core	3.72		4	Core		
1			X					0.6	0.5
	Prepreg	2 x 3313 RC = 57%		3.68	8.6	8.3	Preprea		
8			X					0.6	0.5
	Core	3 x 1652 RC = 50%	core	3.82		18	Core		
9			X				9	0.6	0.5
	Prepreg	2 x 3313 RC = 57%		3.68	8.6	8.3	Prepreg		
10			X				🗖 10	0.6	0.5
	Core	1 x 3313 RC = 55%	core	3.72		4	Core		
11			Х				11	0.6	0.5
	Prepreg	2 x 3313 RC = 57%		3.68	8.6	8.3	Prepreg		
12			Х				<b>1</b> 2	0.6	0.5
	Core	1 x 3313 RC = 55%	core	3.72		4	Core		
13			Х				13	0.6	0.5
-	Prepreg	2 x 3313 RC = 57%		3.68	8.6	8.3	Prepred		
14			X				14	0.6	0.5
	Core	1 x 3313 RC = 55%	core	3.72		4	Core		
15	23.0		X	0.72				0.6	0.5
10	Prepreg	1 x 3313 RC = 57%	~	3.68	43	4 1	Broprog	0.0	0.0
16	, repreg	1 X 3313 1 (0 = 57 %		0.00	т.б	71		22	15
10						0.7	Solder Mack	2.2	1.5
						0.7			
						07.2	110.1	12.0	<u> </u>
						Matorial	110.1		<u> </u>
						Thioksoos	Total Thiskness	Thickness	
						THICKNESS	TOTAL THICKNESS	THICKNESS	
									120

## **Signal Integrity Test Vehicle Highlights**

- Differential Pair lengths from 15" to 60" with a backplane and daughter card configuration
- Reverse Treat Copper (RTF) and VLP-2 Copper used in the same board
- Measurement Opportunities
  - Differential skew
  - Loss tangent
  - Dielectric constant
  - Effect of copper roughness on overall loss
- Amphenol Exceed connectors used
- Any combination of boards/different laminate material can be plugged together to represent classic backplane daughter card configuration



### **S-Parameter Product Comparisons**



#### Courtesy of Speeding Edge

### **Insertion Loss Data**

GHz	Tachyon (VLP2)	Teragreen (VLP2)	I-Tera (VLP2)	Meg6 (HVLP)	Chronon (VLP2)	I-Speed (VLP2)	Gigasync (VLP2)	Meg4 (RTF)
1.25	-0.165	-0.175	-0.169	-0.173	-0.185	-0.179	-0.183	-0.196
2.00	-0.219	-0.231	-0.224	-0.228	-0.245	-0.251	-0.256	-0.278
3.00	-0.271	-0.298	-0.286	-0.293	-0.304	-0.324	-0.341	-0.363
4.00	-0.325	-0.361	-0.351	-0.359	-0.363	-0.406	-0.431	-0.464
5.00	-0.390	-0.433	-0.425	-0.431	-0.424	-0.499	-0.526	-0.570
6.00	-0.445	-0.503	-0.504	-0.504	-0.520	-0.593	-0.640	-0.696
7.00	-0.481	-0.551	-0.553	-0.558	-0.604	-0.679	-0.753	-0.795
8.00	-0.521	-0.604	-0.608	-0.624	-0.653	-0.764	-0.856	-0.895
9.00	-0.570	-0.645	-0.664	-0.678	-0.706	-0.846	-0.956	-1.000
10.00	-0.618	-0.711	-0.730	-0.745	-0.780	-0.938	-1.071	-1.101
11.00	-0.648	-0.758	-0.784	-0.804	-0.861	-1.034	-1.189	-1.193
12.00	-0.684	-0.796	-0.826	-0.869	-0.914	-1.095	-1.271	-1.295
13.00	-0.736	-0.864	-0.891	-0.943	-0.986	-1.194	-1.396	-1.413
14.00	-0.755	-0.909	-0.938	-0.994	-1.048	-1.258	-1.500	-1.510
15.00	-0.795	-0.953	-1.016	-1.091	-1.101	-1.345	-1.641	-1.663
16.00	-0.849	-1.016	-1.060	-1.149	-1.198	-1.523	-1.826	-1.828

### **Skew Data Product Comparison**

	Vertical	Skew on 9"	Line (ps)	Horizontal Skew on 14" Line (ps)			
Products	Minimum	Maximum	Average	Minimum	Maximum	Average	
IS415	0	8	5	30	123	88	
I-SPEED	3	10	4.5	1	59	18	
I-TERA IS	1	4	2.5	4	59	24.6	
FR408HR	1	8	5	3	43	20	
MEG 6	0	4	2	2	37	13	
MEG 4	1	2	1	4	28	13	
I-TERA #2	0	1	2	0	23	8	
FR408HR IS	0	7	4.6	6	20	11.8	
I-TERA #1	1	12	6	1	13	9.5	
I-SPEED IS	1	4	1.3	5	12	7.5	
I-SPEED #2	2	4.5	2.8	0	11	3.7	
TerraGreen	0	5	3	2	9	5	
I-SPEED IS #2	0	3	0.9	0	8.5	3.1	
Tachyon	2	4	3	1	8	4	
Chronon	0	2	1	1	5	3	
GigaSync #1	0	4	1	0	4	2	
GigaSync #2	0	2	1	0	3	2	

# Conclusions

- Isola has two high performance product offerings for skew mitigation
  - Tachyon
  - Chronon
- Products offer process compatibility with Isola low-cost materials in hybrid constructions
- Availability is immediate and materials have been sampled and thoroughly tested
  - Alcatel MRT5
  - Cisco SI TV



# References

Y. Shlepnev, C.Nwachukwu, "Modelling Skew and Jitter induced by Fiber weave effect in PCB dielectrics", IEEE International Symposium on Electromagnetic Compatibility, Raleigh, North Carolina, August 2014.





# **Thank You!**